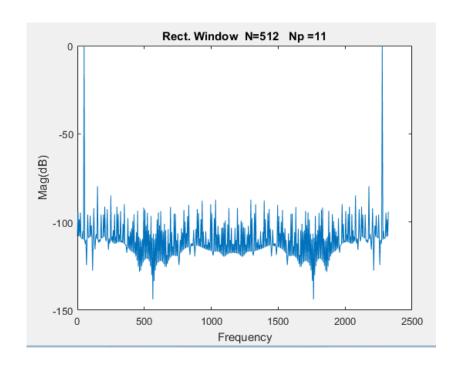
EE 505 Lecture 7

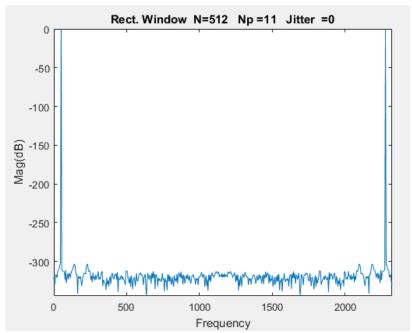
Windowing Spectral Performance of Data Converters

- Time Quantization
- Amplitude Quantization

Clock Jitter Statistical Circuit Modeling

MatLab comparison: 512 Samples with Standard Sweep

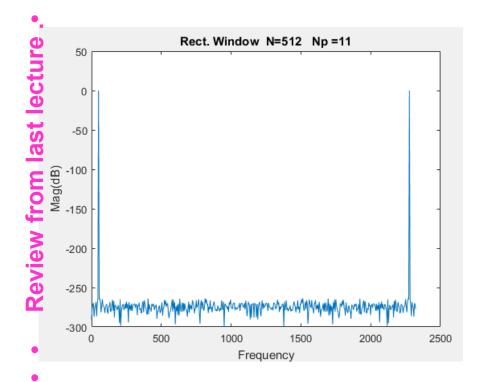


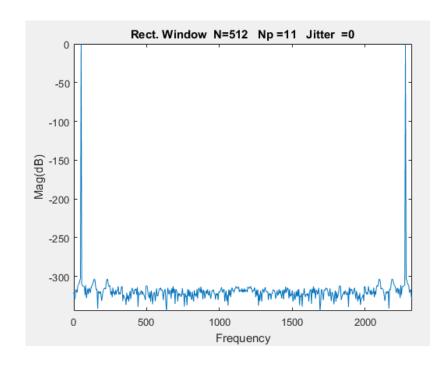


Spectre Results

MatLab Results

MatLab comparison: 512 Samples with Strobe Period Sweep





Spectre Results

MatLab Results

Considerations for Spectral Characterization

- Tool Validation
- FFT Length
- Importance of Satisfying Hypothesis
 - NP is an integer
 - Band-limited excitation



Windowing - a strategy to address the problem of requiring precisely an integral number of periods to use the DFT for Spectral analysis?

- Windowing is sometimes used
- Windowing is sometimes misused

Example

WLOG assume f_{SIG}=50Hz

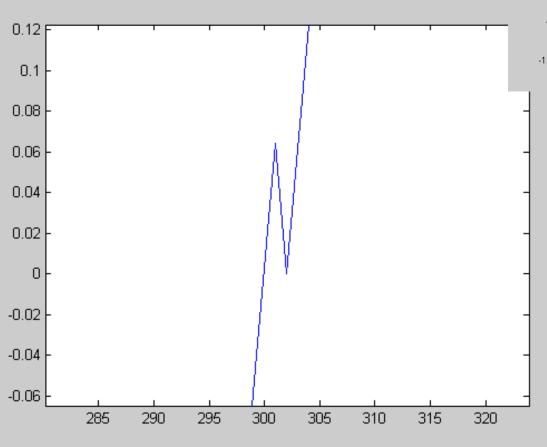
$$V_{IN} = \sin(\omega t) + 0.5\sin(2\omega t)$$

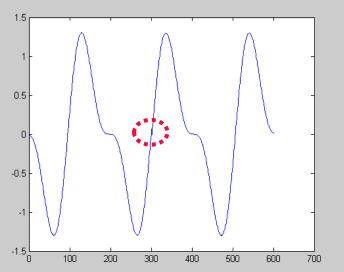
$$\omega = 2\pi f_{SIG}$$

Consider $N_p=20.01$ N=4096

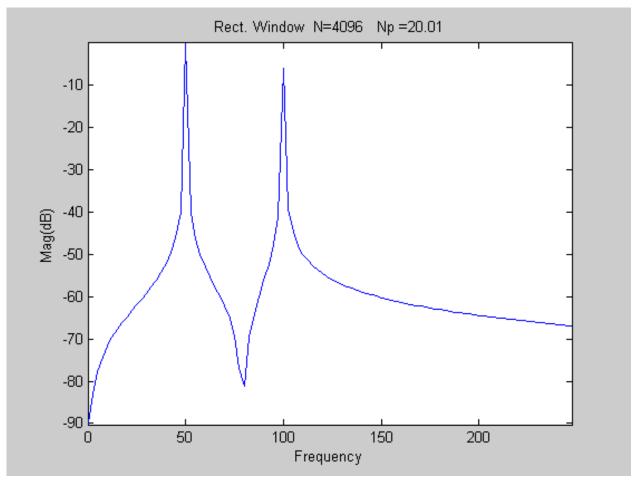
Deviation from hypothesis is .05% of the sampling window

Input Waveform

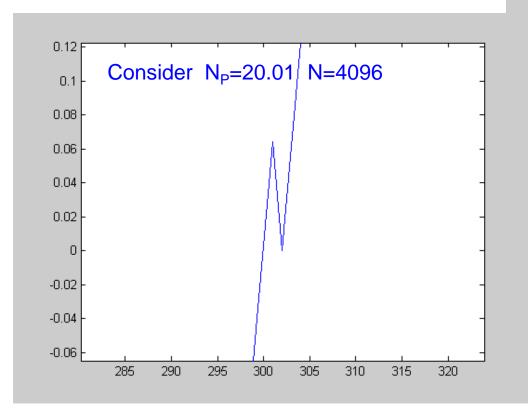


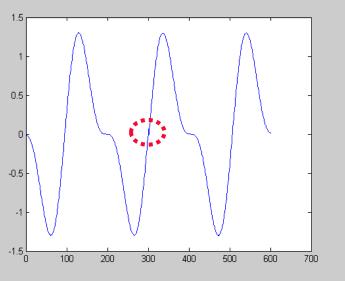


Spectral Response with Non-Coherent Sampling



(zoomed in around fundamental)





Even with $N_p=20.001$ had significant degradation

Extremely small discontinuity associated with non-coherent sampling causes Significant degradations in spectral response if DFT (and Theorem) used

Windowing

Windowing is the weighting of the time domain function to maintain continuity at the end points of the sample window

Well-studied window functions:

- Rectangular (also with appended zeros)
- Triangular
- Hamming
- Hanning
- Blackman

Sometimes termed a boxcar window

Uniform weight

Can append zeros

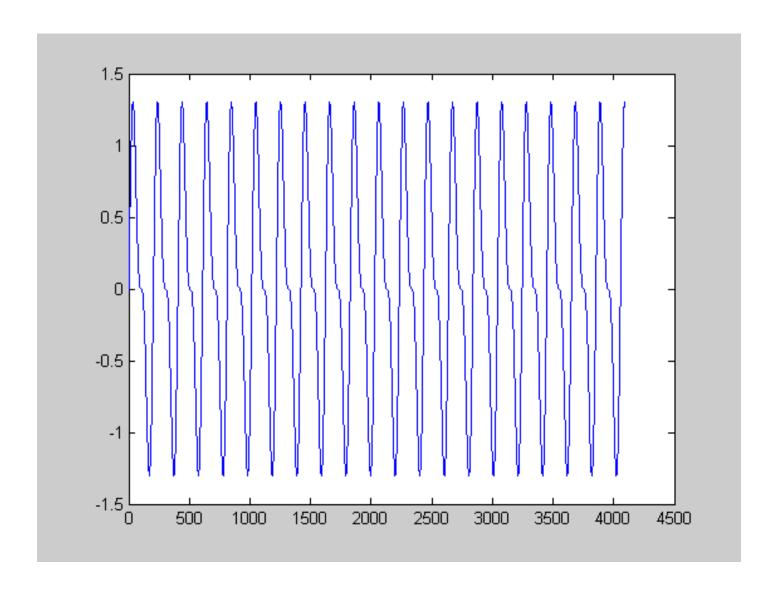
Without appending zeros equivalent to no window

Assume f_{SIG}=50Hz

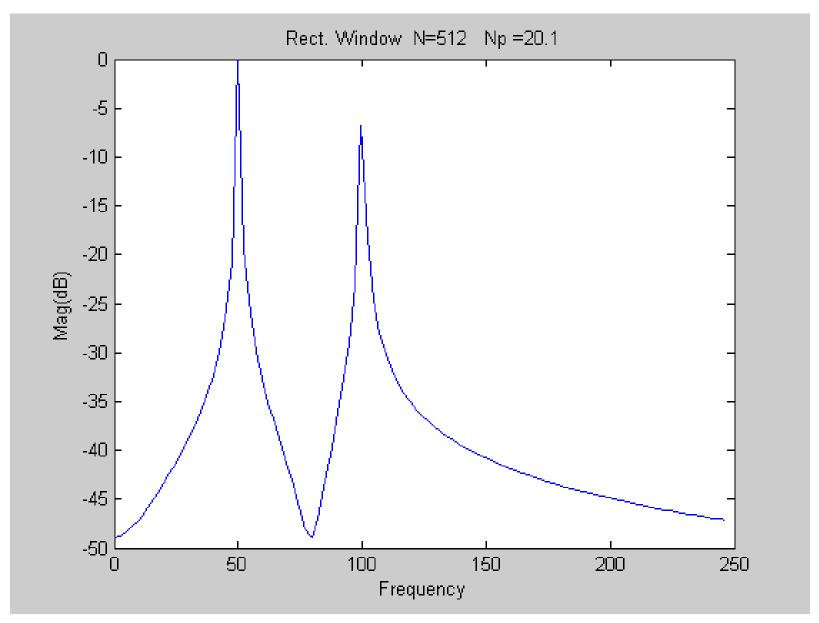
$$V_{IN} = \sin(\omega t) + 0.5\sin(2\omega t)$$

$$\omega = 2\pi f_{SIG}$$

Consider $N_p=20.1$ N=512



Spectral Response with Non-coherent sampling



(zoomed in around fundamental)

Columns 1 through 7

-48.8444 -48.7188 -48.3569 -47.7963 -47.0835 -46.2613 -45.3620

Columns 8 through 14

-44.4065 -43.4052 -42.3602 -41.2670 -40.1146 -38.8851 -37.5520

Columns 15 through 21

-36.0756 -34.3940 -32.4043 -29.9158 -26.5087 -20.9064 -0.1352

Columns 22 through 28

-19.3242 -25.9731 -29.8688 -32.7423 -35.1205 -37.2500 -39.2831

Columns 29 through 35

-41.3375 -43.5152 -45.8626 -48.0945 -48.8606 -46.9417 -43.7344

Columns 1 through 7

-48.8444 -48.7188 -48.3569 -47.7963 -47.0835 -46.2613 -45.3620

Columns 8 through 14

-44.4065 -43.4052 -42.3602 -41.2670 -40.1146 -38.8851 -37.5520

Columns 15 through 21

-36.0756 -34.3940 -32.4043 -29.9158 -26.5087 -20.9064 -0.1352

Columns 22 through 28

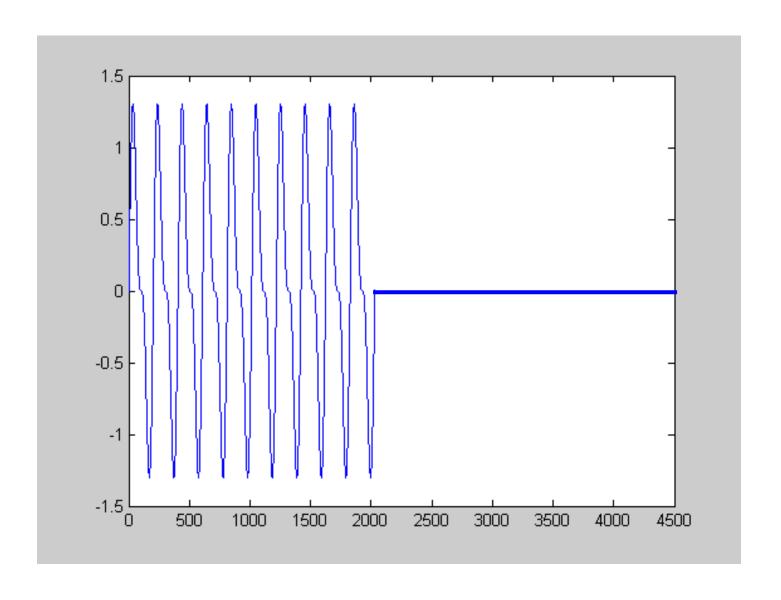
-19.3242 -25.9731 -29.8688 -32.7423 -35.1205 -37.2500 -39.2831

Columns 29 through 35

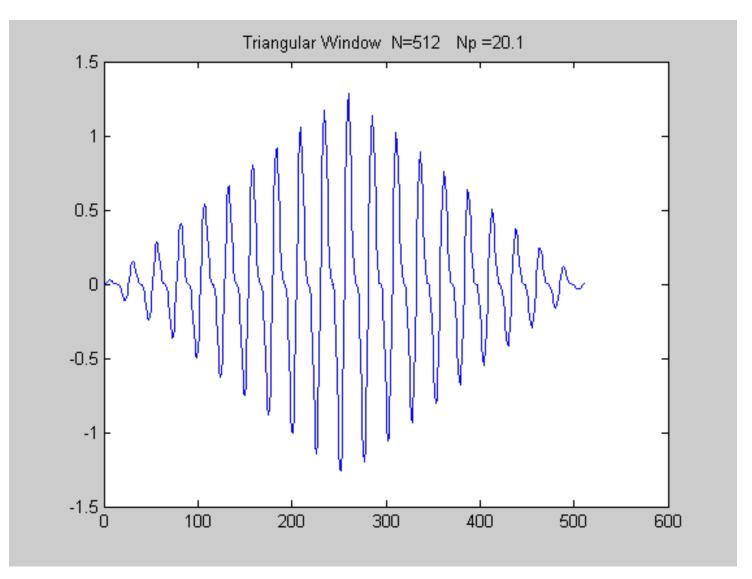
-41.3375 -43.5152 -45.8626 -48.0945 -48.8606 -46.9417 -43.7344

Energy spread over several frequency components

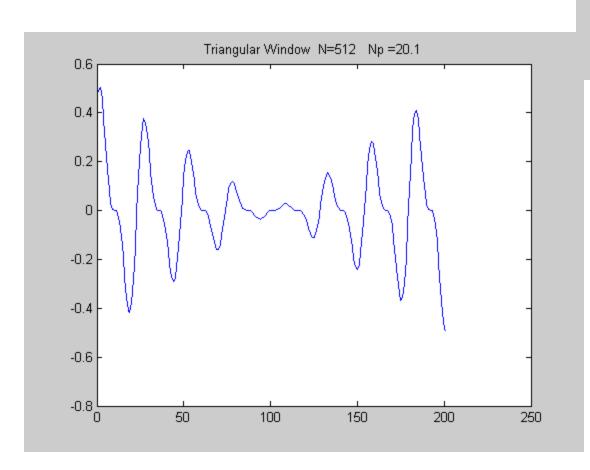
Rectangular Window (with appended zeros)

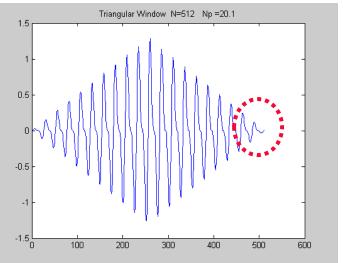


Triangular Window

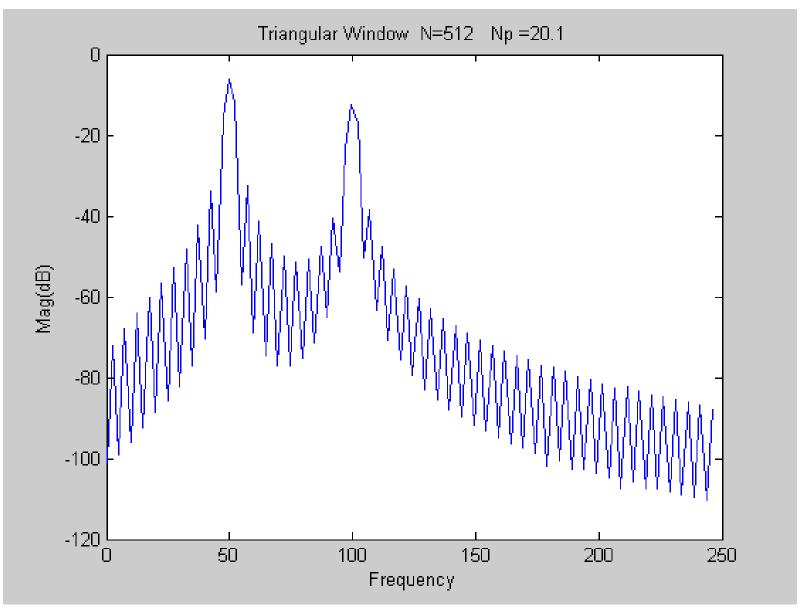


Triangular Window



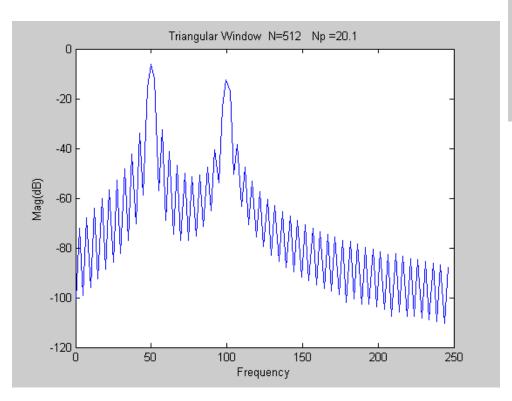


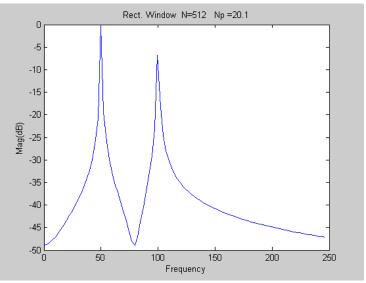
Spectral Response with Non-Coherent Sampling and Windowing



(zoomed in around fundamental)

Triangular Window





Triangular Window

Columns 1 through 7

-100.8530 -72.0528 -99.1401 -68.0110 -95.8741 -63.9944 -92.5170

Columns 8 through 14

-60.3216 -88.7000 -56.7717 -85.8679 -52.8256 -82.1689 -48.3134

Columns 15 through 21

-77.0594 -42.4247 -70.3128 -33.7318 -58.8762 -15.7333

-6.0918

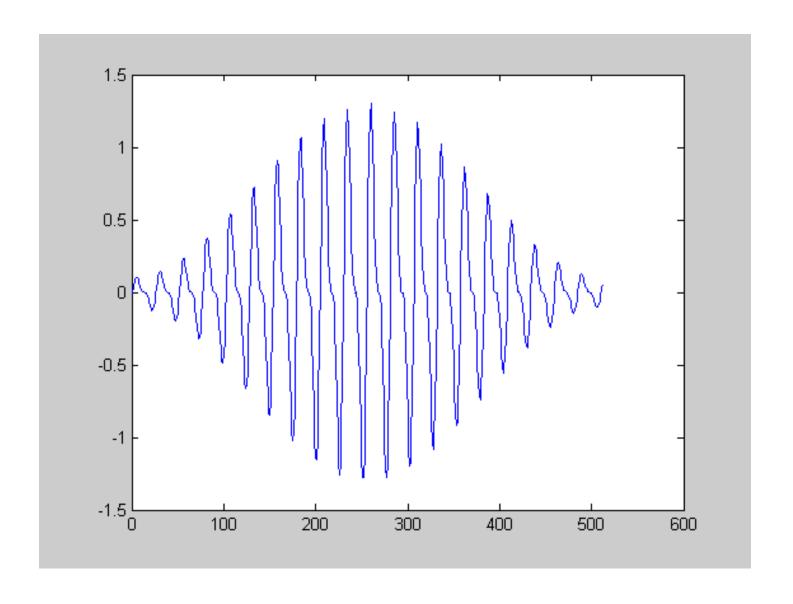
Columns 22 through 28

-12.2463 -57.0917 -32.5077 -68.9492 -41.3993 -74.6234 -46.8037

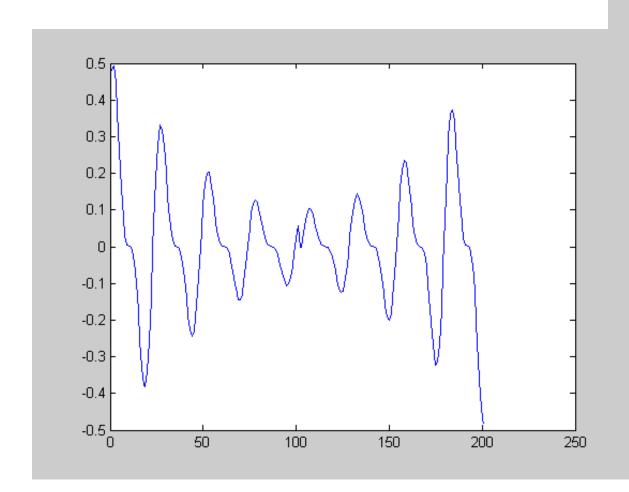
Columns 29 through 35

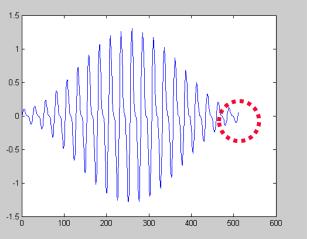
-77.0686 -50.1054 -77.0980 -51.5317 -75.1218 -50.8522 -71.2410

Hamming Window

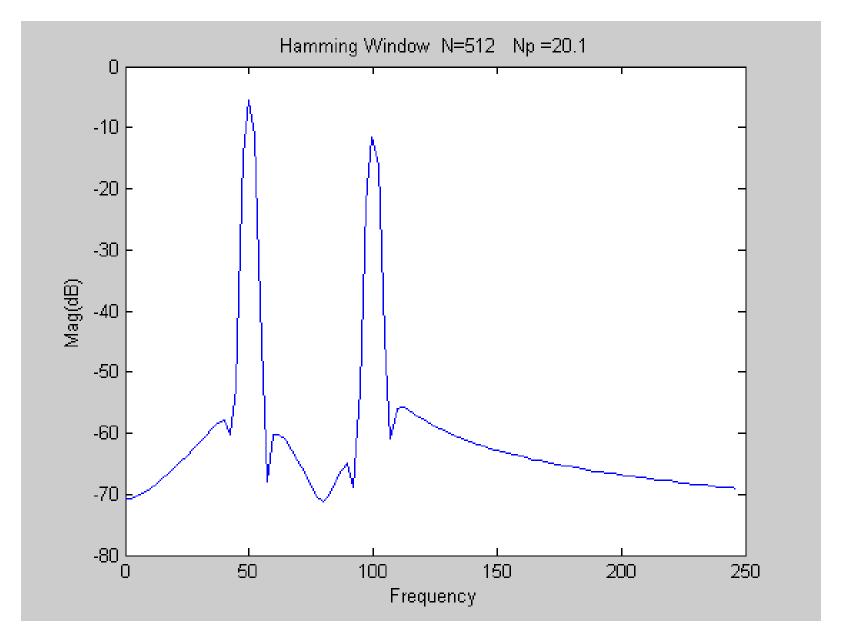


Hamming Window



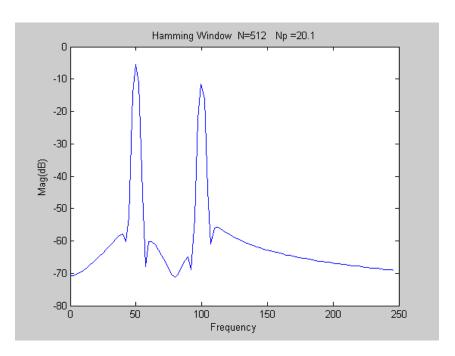


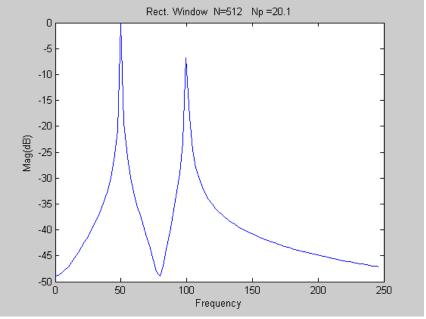
Spectral Response with Non-Coherent Sampling and Windowing



(zoomed in around fundamental)

Comparison with Rectangular Window





Hamming Window

Columns 1 through 7

-70.8278 -70.6955 -70.3703 -69.8555 -69.1502 -68.3632 -67.5133

Columns 8 through 14

-66.5945 -65.6321 -64.6276 -63.6635 -62.6204 -61.5590 -60.4199

Columns 15 through 21

-59.3204 -58.3582 -57.8735 -60.2994 -52.6273 -14.4702

-5.4343

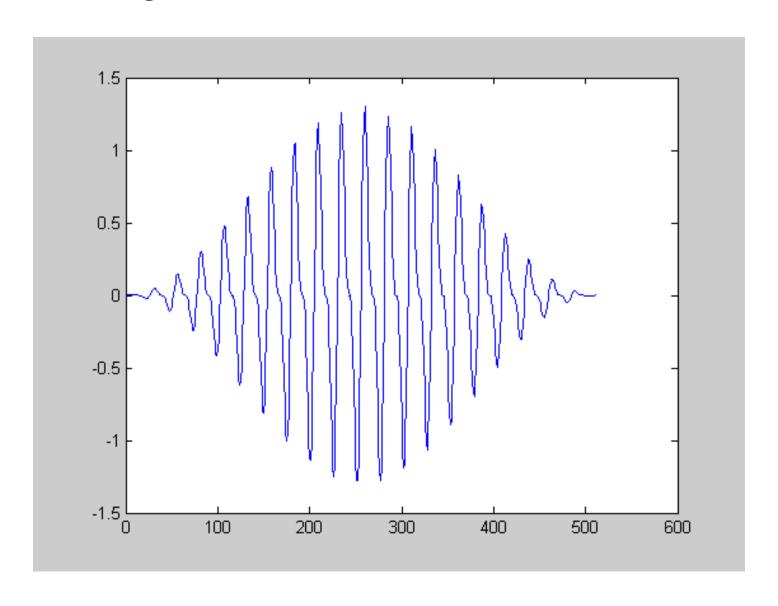
Columns 22 through 28

-11.2659 -45.2190 -67.9926 -60.1662 -60.1710 -61.2796 -62.7277

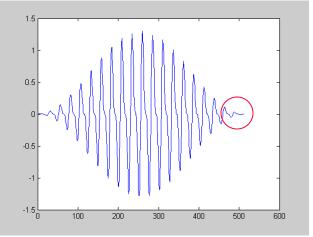
Columns 29 through 35

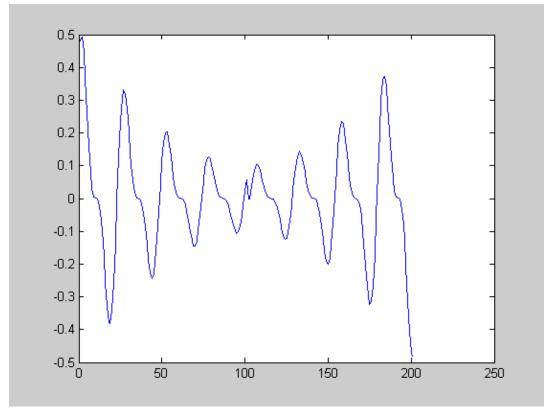
-64.3642 -66.2048 -68.2460 -70.1835 -71.1529 -70.2800 -68.1145

Hanning Window

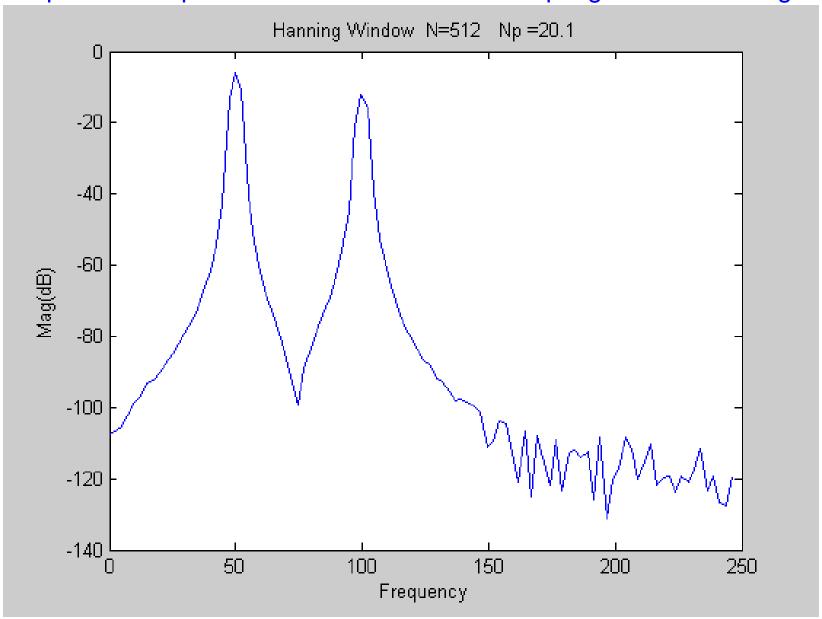


Hanning Window



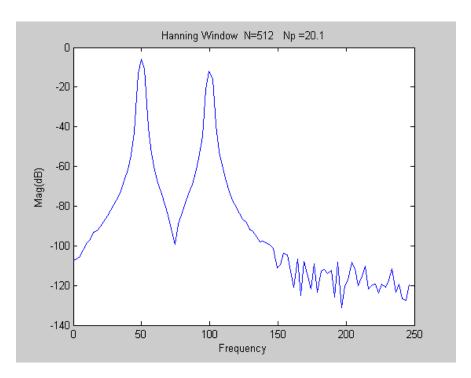


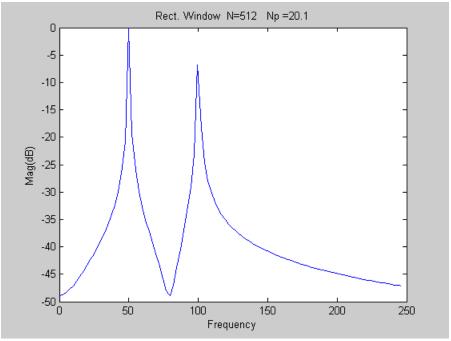
Spectral Response with Non-Coherent Sampling and Windowing



(zoomed in around fundamental)

Comparison with Rectangular Window





Hanning Window

Columns 1 through 7

-107.3123 -106.7939 -105.3421 -101.9488 -98.3043 -96.6522 -93.0343

Columns 8 through 14

-92.4519 -90.4372 -87.7977 -84.9554 -81.8956 -79.3520 -75.8944

Columns 15 through 21

-72.0479 -67.4602 -61.7543 -54.2042 -42.9597 -13.4511 (-6.0601

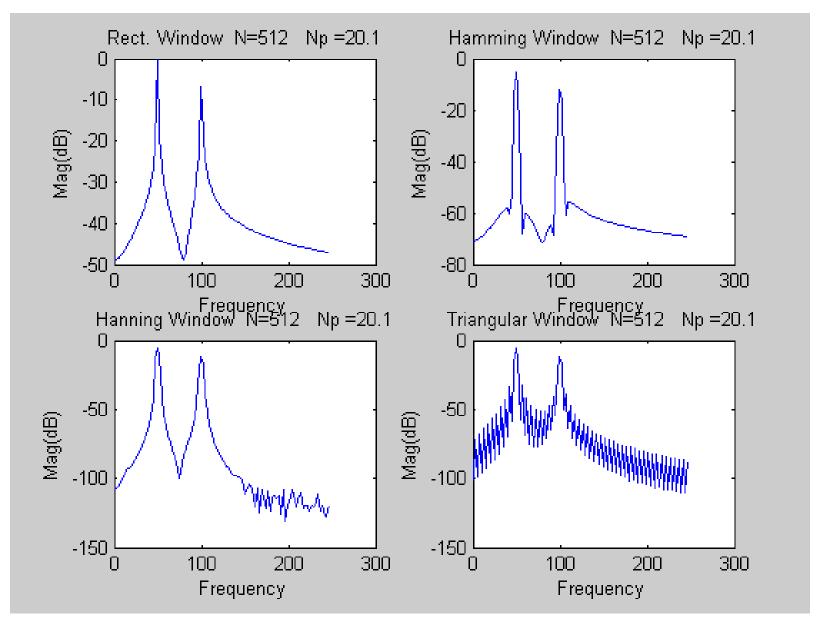
Columns 22 through 28

-10.8267 -40.4480 -53.3906 -61.8561 -68.3601 -73.9966 -79.0757

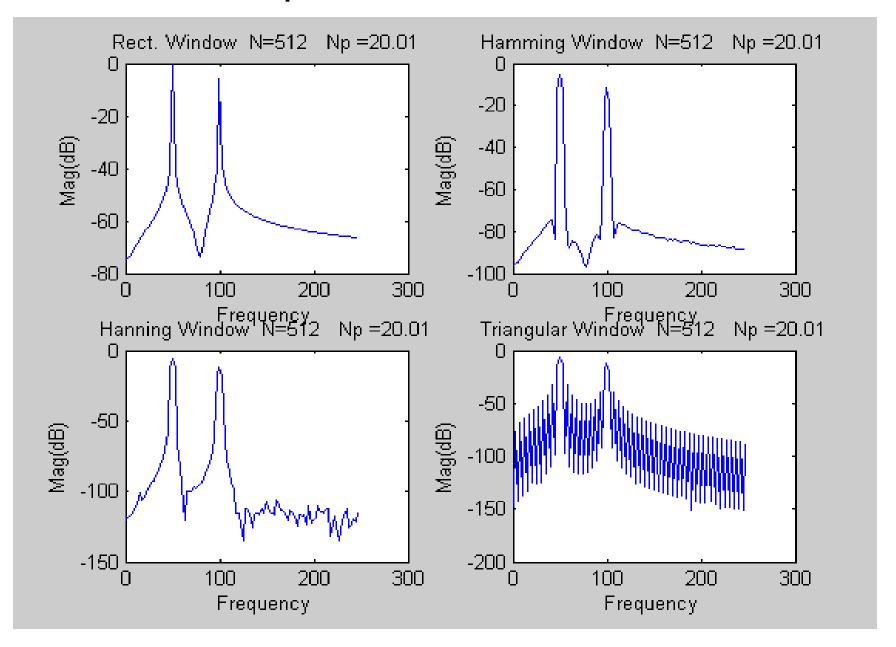
Columns 29 through 35

-84.4318 -92.7280 -99.4046 -89.0799 -83.4211 -78.5955 -73.9788

Comparison of 4 windows



Comparison of 4 windows

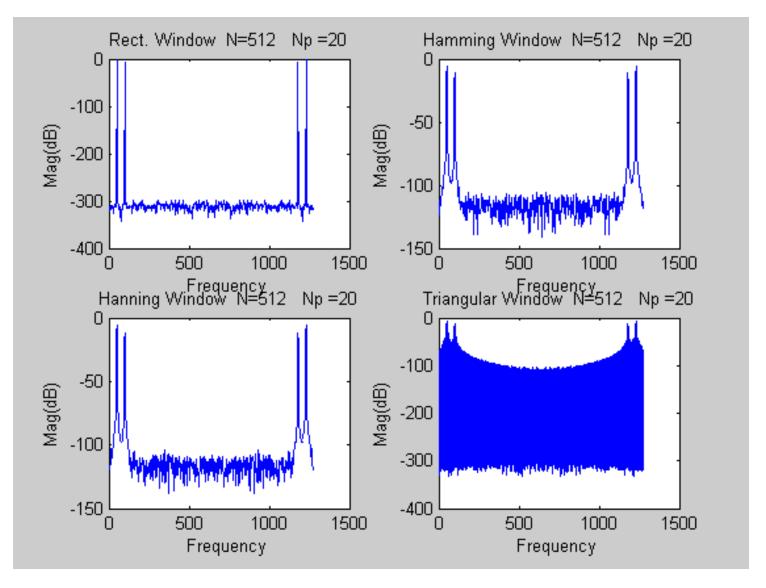


Preliminary Observations about Windows

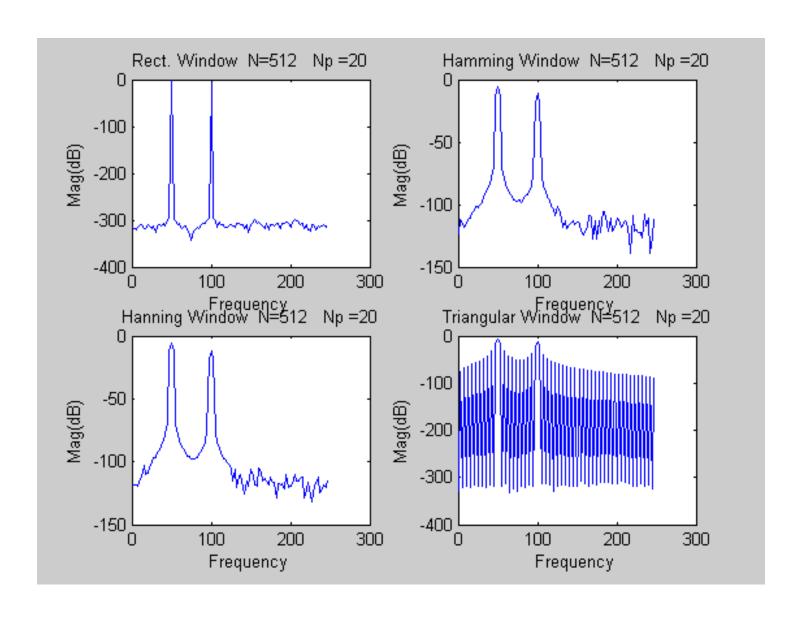
- Provide separation of spectral components
- Energy can be accumulated around spectral components
- Simple to apply
- Some windows work much better than others

But – windows do not provide dramatic improvement and ...

Comparison of 4 windows when sampling hypothesis are satisfied



Comparison of 4 windows



Preliminary Observations about Windows

- Provide separation of spectral components
- Energy can be accumulated around spectral components
- Simple to apply
- Some windows work much better than others

But – windows do not provide dramatic improvement and can significantly degrade performance if sampling hypothesis are met

Addressing Spectral Analysis Challenges

- Problem Awareness
- Windowing and Filtering



Post-processing

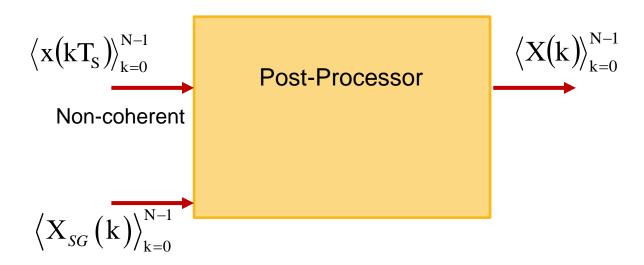
Method of circumventing the coherent sampling problem

Can also be used for addressing spectral purity problem for test signal generation



- Easily implemented in MATLAB
- Will be considered in the laboratory
- "Removes" fundamental from samples and replaces with coherent fundamental before taking DFT

Post-processing



- Easily implemented in MATLAB
- Will be considered in the laboratory
- "Removes" fundamental from samples and replaces with coherent fundamental before taking DFT
- Removes spectral impurity of input test signal generator when testing data converters

Issues of Concern for Spectral Analysis

An integral number of periods is critical for spectral analysis

Not easy to satisfy this requirement in the laboratory

Windowing can help but can hurt as well

Out of band energy can be reflected back into bands of interest

Characterization of CAD tool environment is essential

Spectral Characterization of high-resolution data converters requires particularly critical consideration to avoid simulations or measurements from masking real performance

Spectral Characterization of Data Converters

- Distortion Analysis
- Time Quantization Effects
 - of DACs
 - of ADCs
- Amplitude Quantization Effects
 - of DACs
 - of ADCs
- Clock Jitter

time and amplitude depicted

Zero-order sample/hold on DAC or zero-order hold on ADC interpreted output

DAC Assume DAC will be used to generate a continuous time signal

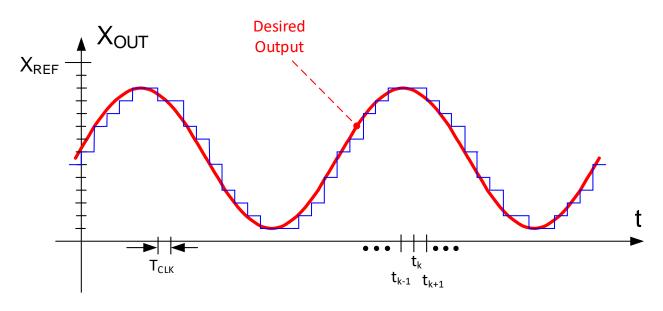
Assume DAC is driven by a clock of period T_{CLK}

DAC inputs will be a discrete sequence $\vec{X}(t_k) = \langle x_{quant}(t_k) \rangle$

DAC inputs can change only at times t_k

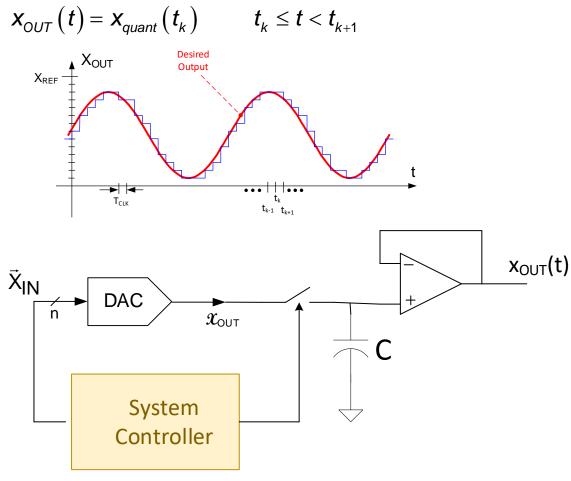
The duration of each DAC input depends upon system

With zero-order S/H, it is assumed that the DAC output remains constant between transaction times $X_{OUT}(t) = X_{quant}(t_k)$ $t_k \le t < t_{k+1}$



time and amplitude depicted
Zero-order sample/hold on DAC or zero-order hold on ADC interpreted output

DAC



Zero-order S/H

Transition points not necessarily uniformly spaced but will assume so in what follows

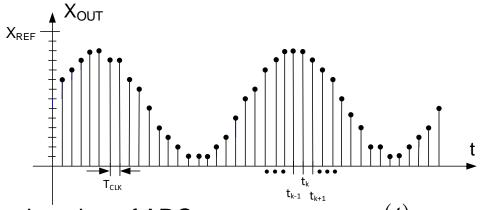
time and amplitude depicted
Zero-order sample/hold on DAC or zero-order hold on ADC intern

Zero-order sample/hold on DAC or zero-order hold on ADC interpreted output

ADC Output is dimensionless sequence

$$\vec{X}(k) = \langle X_{qant}(t_k) \rangle$$

Interpreted output can be represented as a stem plot



Zero-order continuation of ADC output

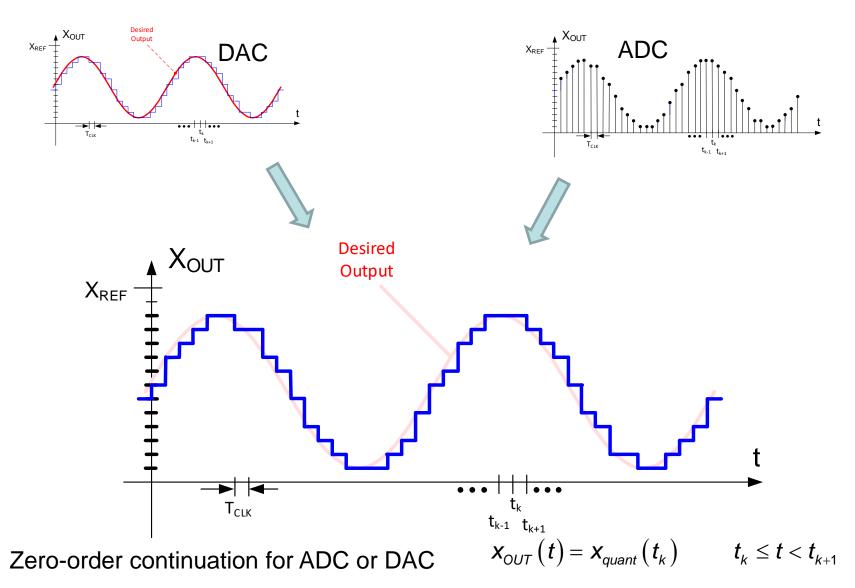
 T_{CLK}

 $_{A}$ X_{OUT}

 X_{REF}

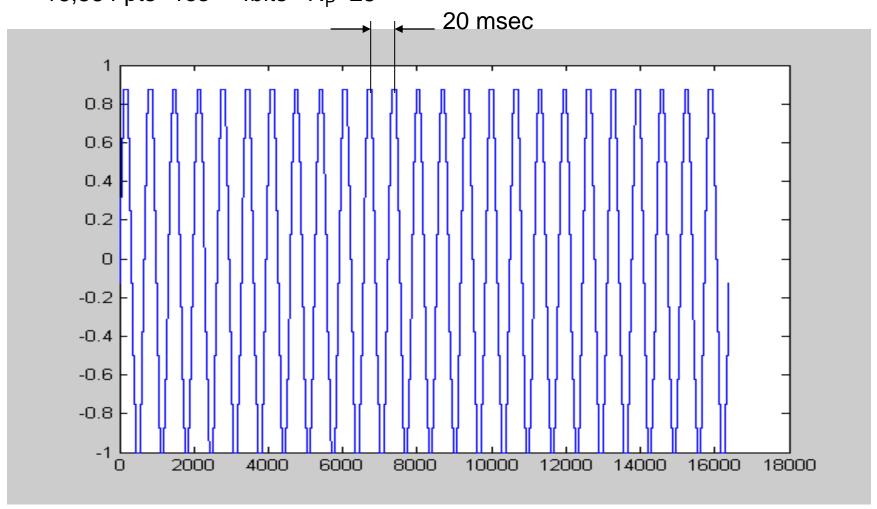
f ADC output
$$X_{OUT}(t) = X_{quant}(t_k)$$
 $t_k \le t < t_{k+1}$ Continuation of Interpreted Input t_k

time and amplitude depicted Zero-order sample/hold on DAC or zero-order hold on ADC interpreted output



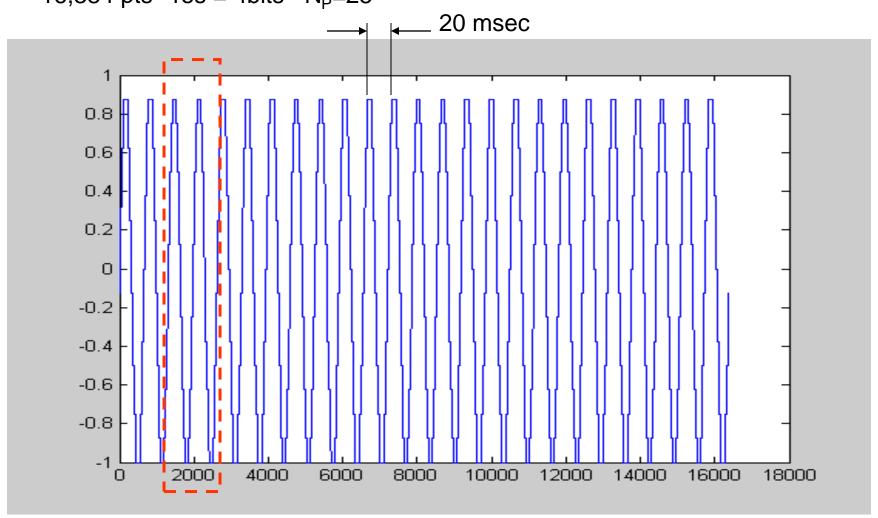
time and amplitude depicted
For zero-order sample/hold on DAC or zero-order hold on ADC interpreted output

16,384 pts res = 4bits $N_p=25$



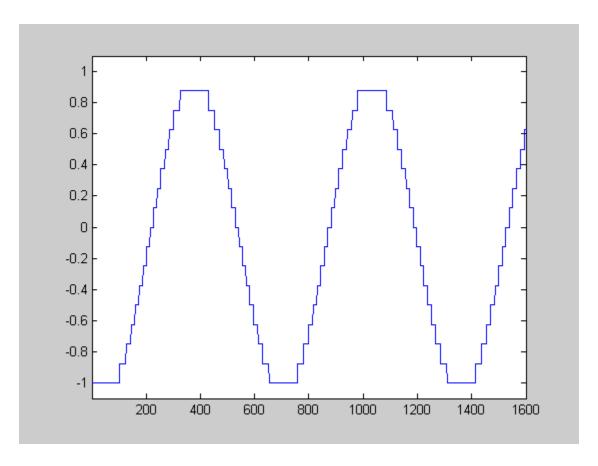
(time and amplitude depicted)

16,384 pts res = 4bits $N_p=25$



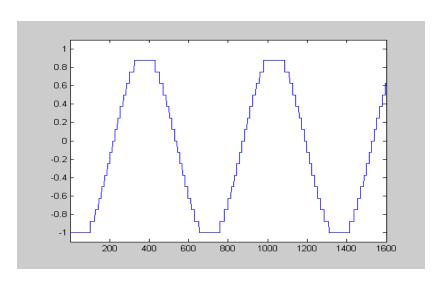
(time and amplitude depicted)

16,384 pts res = 4bits



Is this signal band limited?

(time and amplitude depicted)



Simulation environment:

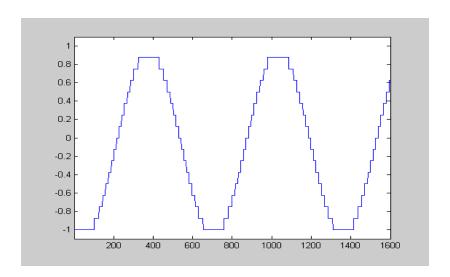
 $N_P=23$ $f_{SIG}=50Hz$ V_{REF} : -1V, 1V

Res: will be varied N=2ⁿ will be varied

Spectral Characterization of Data Converters

- Distortion Analysis
- Time Quantization Effects
 - of DACs
 - of ADCs
- Amplitude Quantization Effects
 - of DACs
 - of ADCs
 - Clock Jitter

(time and amplitude depicted)

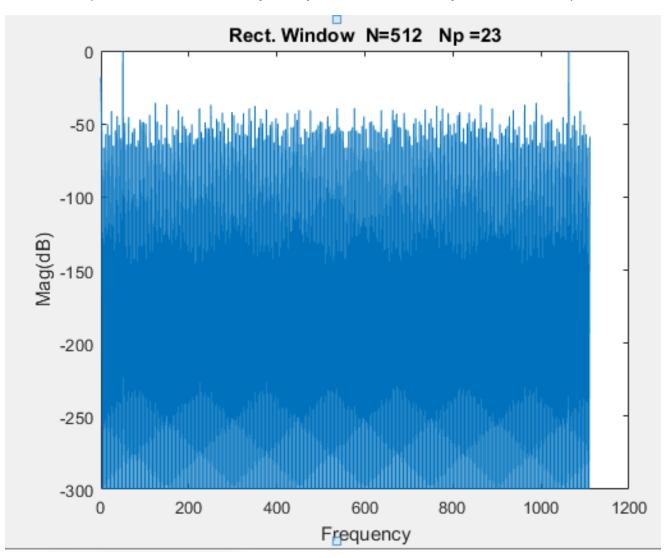


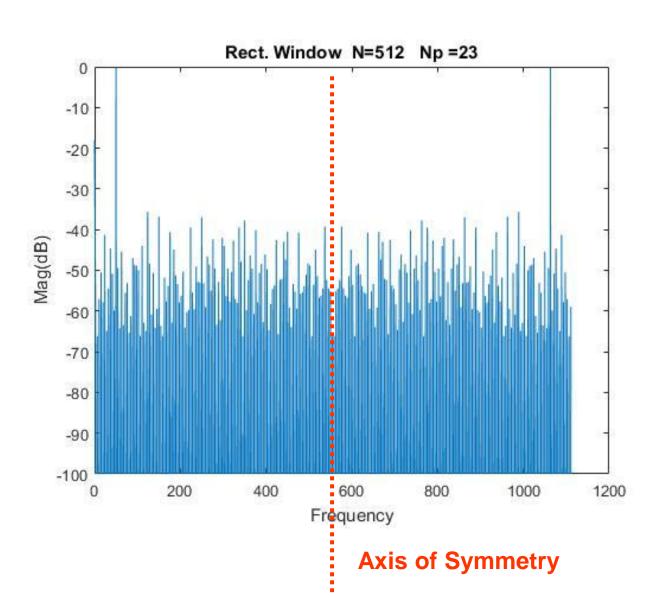
For amplitude quantization, what appear to be horizontal steps in the above figure are not the same

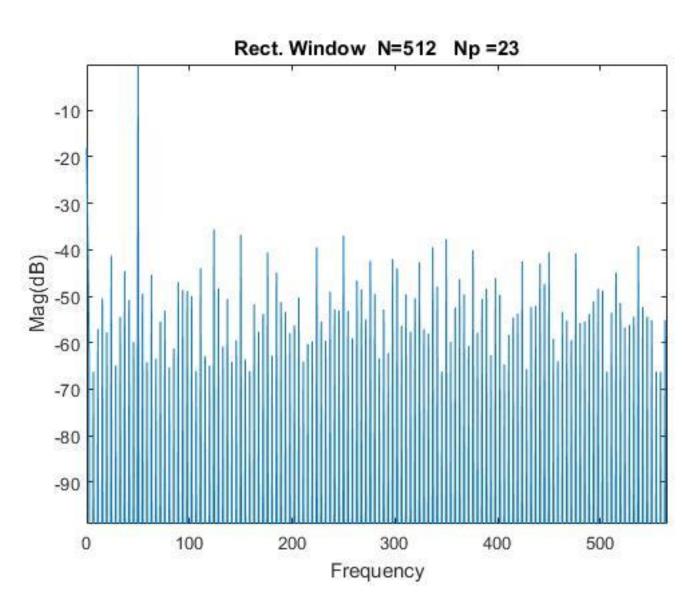
(amplitude quantization not depicted)

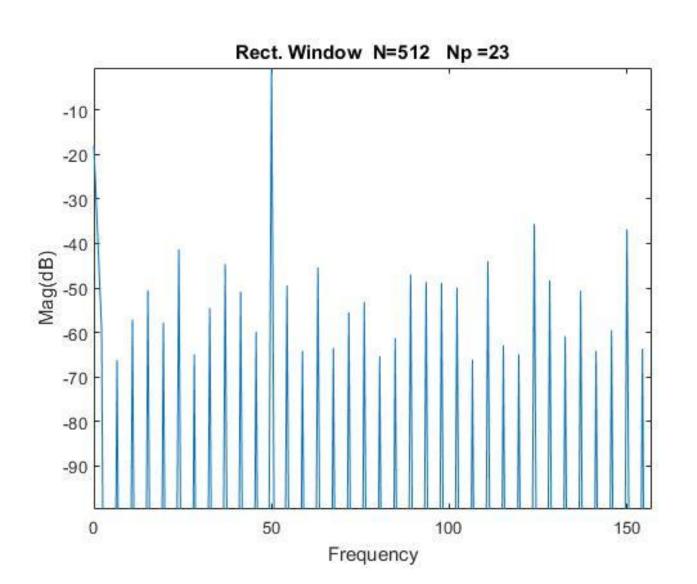
Res = 4 bits

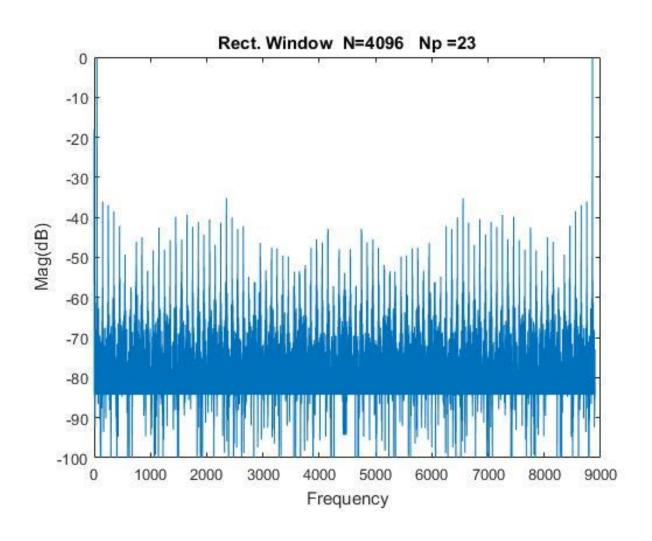
(At each time sample, quantize the amplitude value)

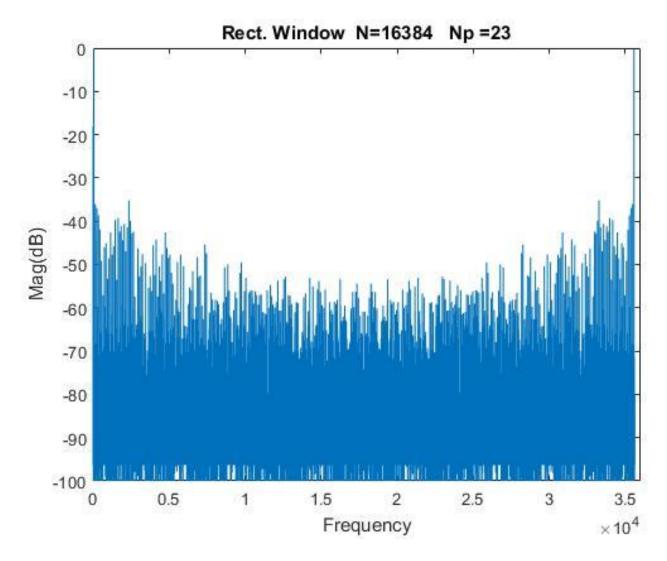






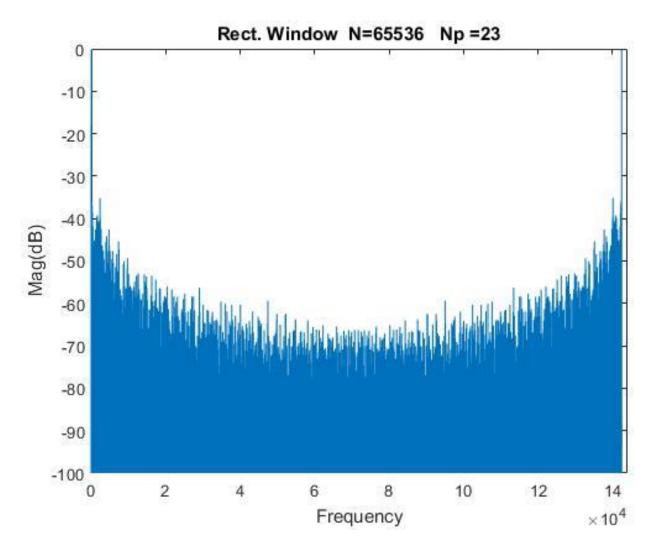




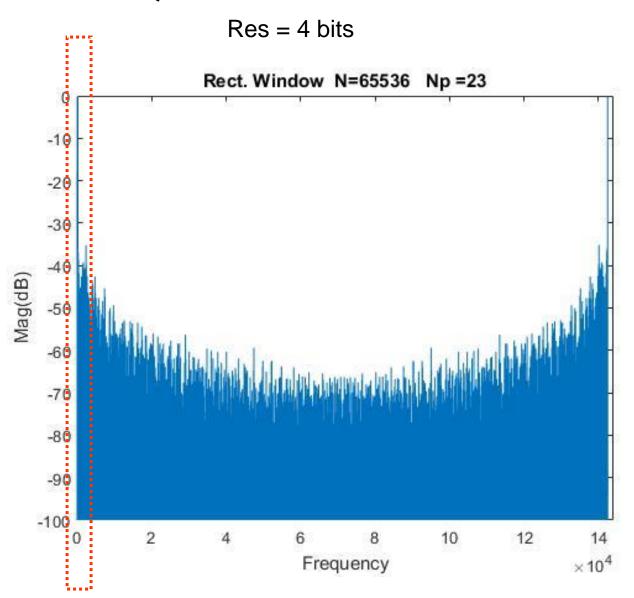


Expect quantization noise effects to be uniformly distributed!!

Res = 4 bits

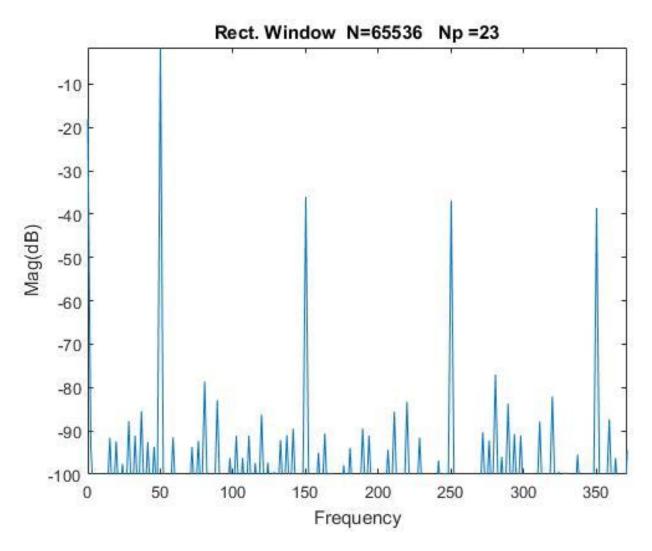


Expect quantization noise effects to be uniformly distributed!!



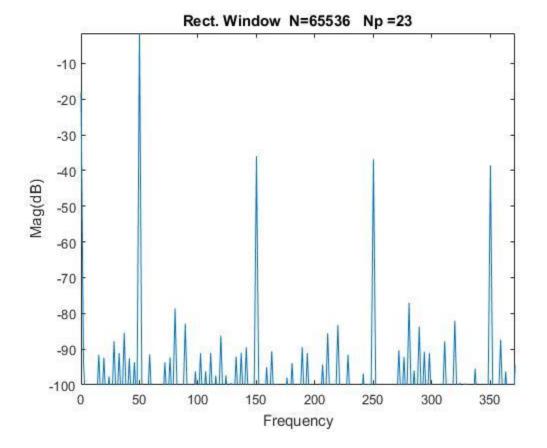
Expect quantization noise effects to be uniformly distributed!!

Res = 4 bits



Note presence of odd-ordered harmonic terms!!

Res = 4 bits

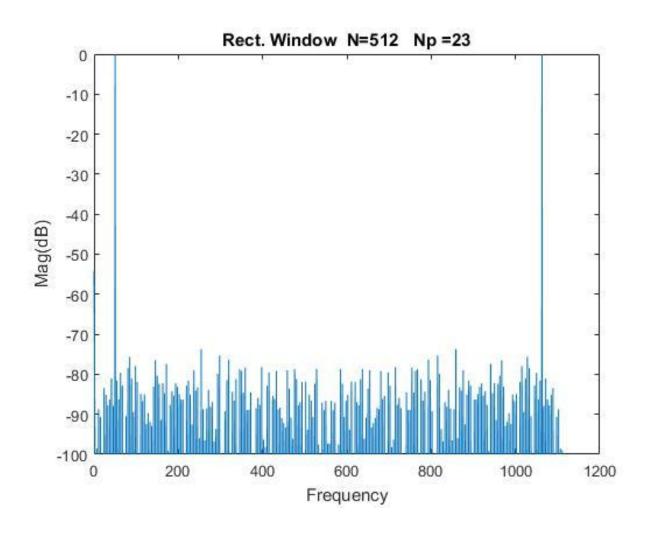


Res (n)	SNR _{corr}	SNR
1	3.86	7.78
2	12.06	13.8
3	19.0	19.82
4	25.44	25.84
5	31.66	31.86
6	37.79	37.88
8	49.90	49.92
10	61.95	61.96

Why are there spectral components present in the quantization noise?

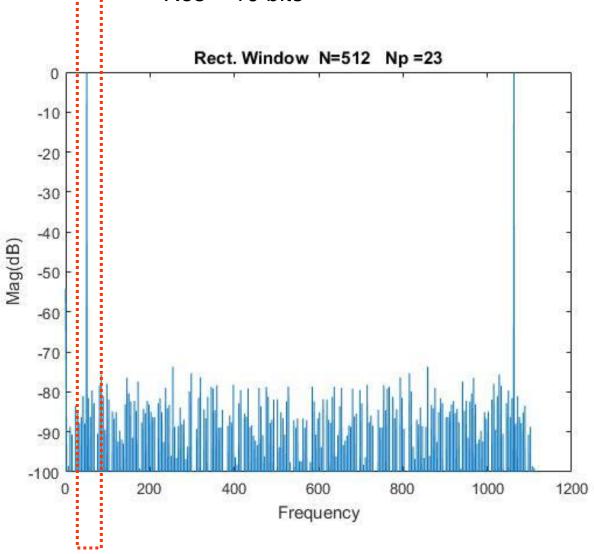
Recall the uncorrelated assumption was good only for about 4 bits or more!

Res = 10 bits

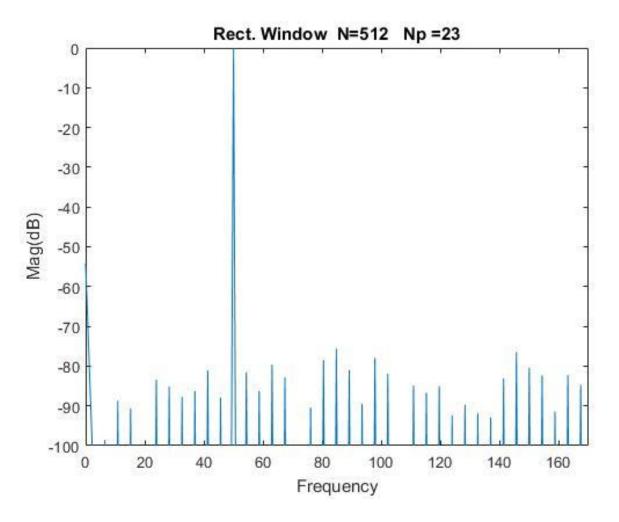


Quantization noise is much more uniform

Res = 10 bits

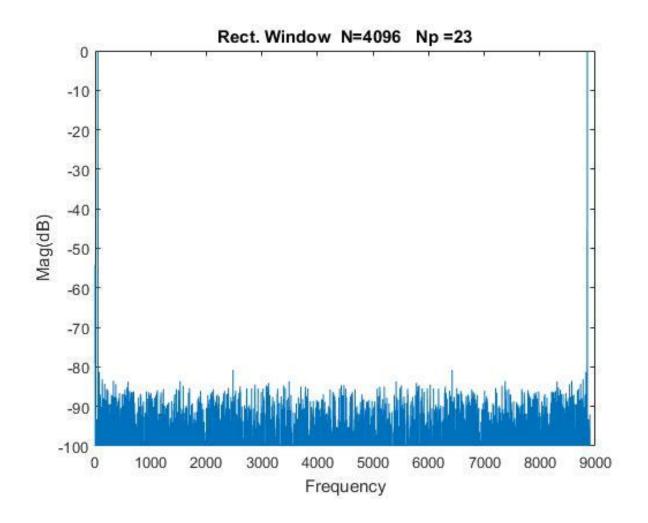


Res = 10 bits



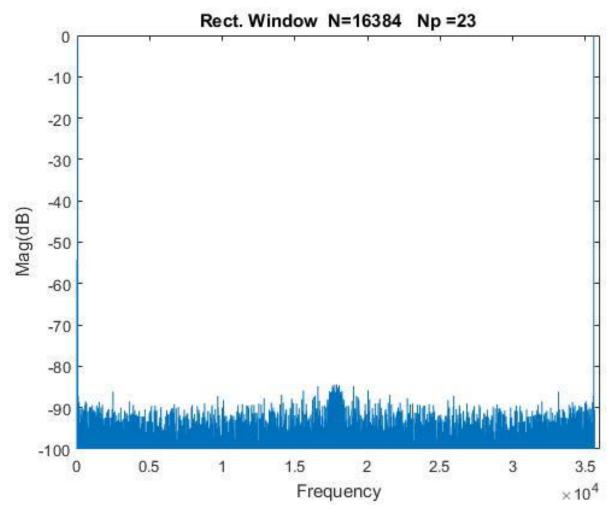
Harmonic Components not Visible

Res = 10 bits



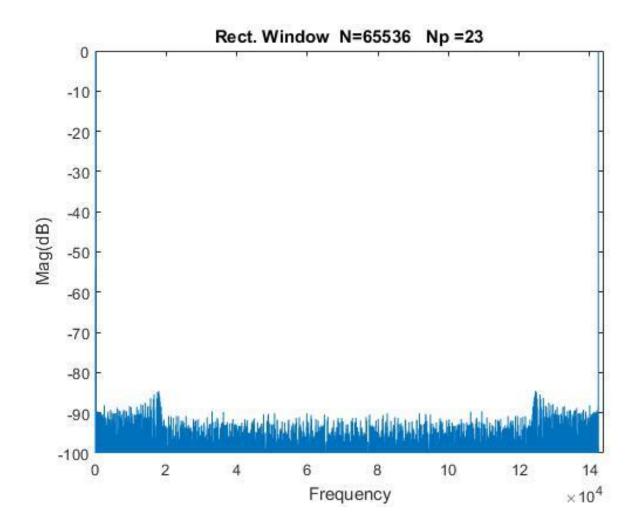
Compared to the previous slide, it appears that the quantization noise has gone down – why does this occur?

Res = 10 bits



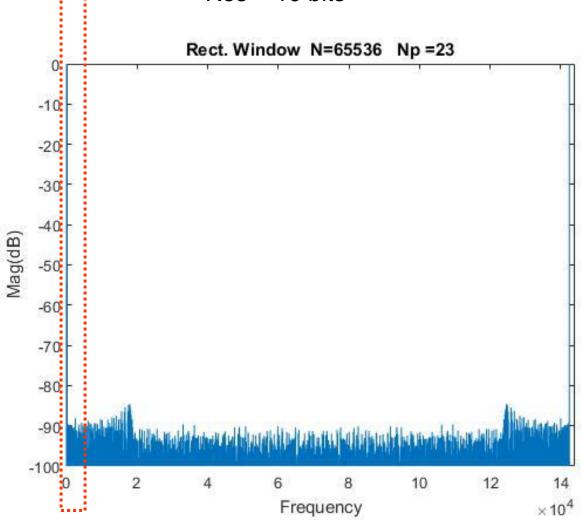
Compared to the previous slide, it appears that the quantization noise has gone down even more – why does this occur?

Res = 10 bits

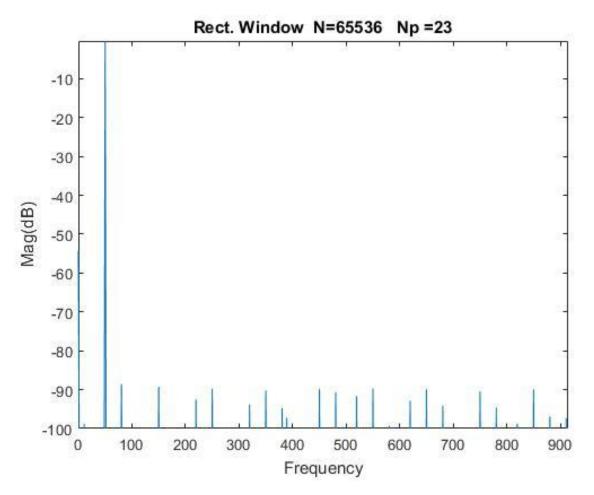


Compared to the previous slides, it appears that the quantization noise has gone down even more – why does this occur?

Res = 10 bits



Res = 10 bits



Very small third harmonic component but does not extend above other noise terms

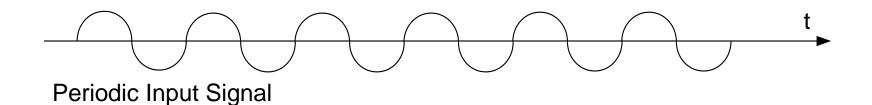
Spectral Characterization

- Amplitude Quantization
 - Does not introduce substantive spectral components for n large
 - Nearly uniformly distributed
 - Decreases with increasing N

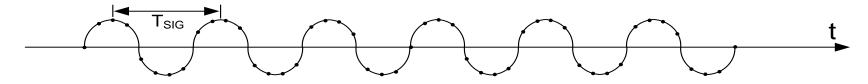
Spectral Characterization of Data Converters

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 - of DACs
 - of ADCs
 - Amplitude Quantization Effects
 - of DACs
 - of ADCs
 - Clock Jitter

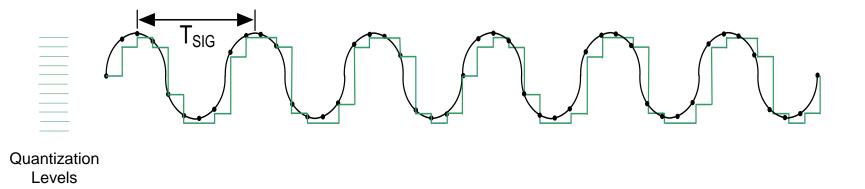
Spectral Characteristics of DACs and ADCs

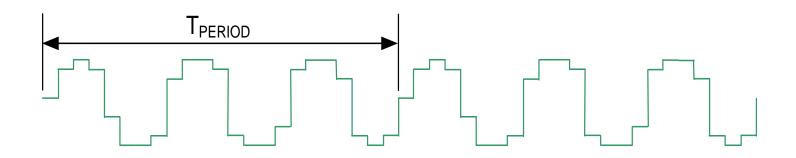


Sampling Clock

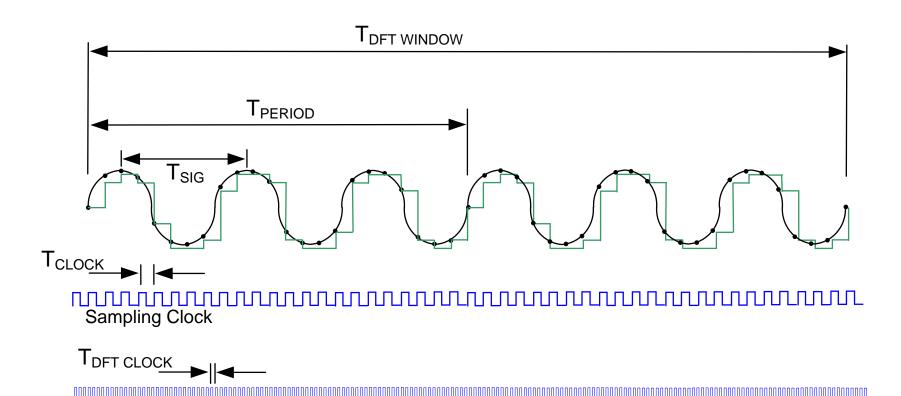


Sampled Input Signal (showing time points where samples taken)

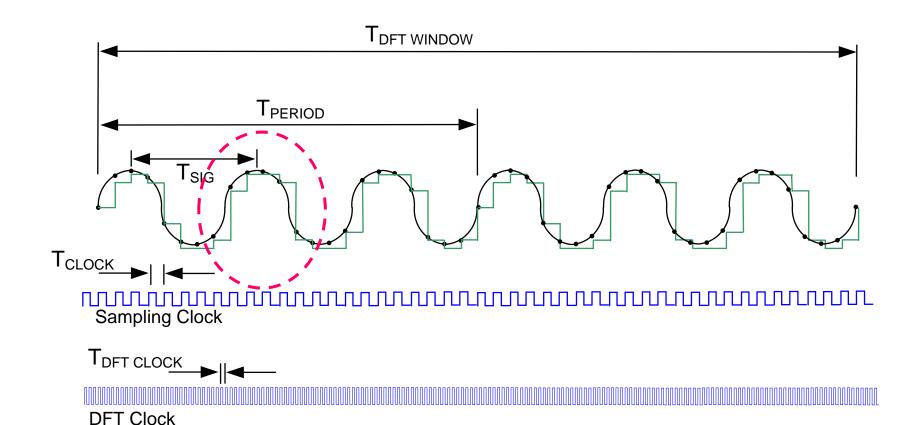


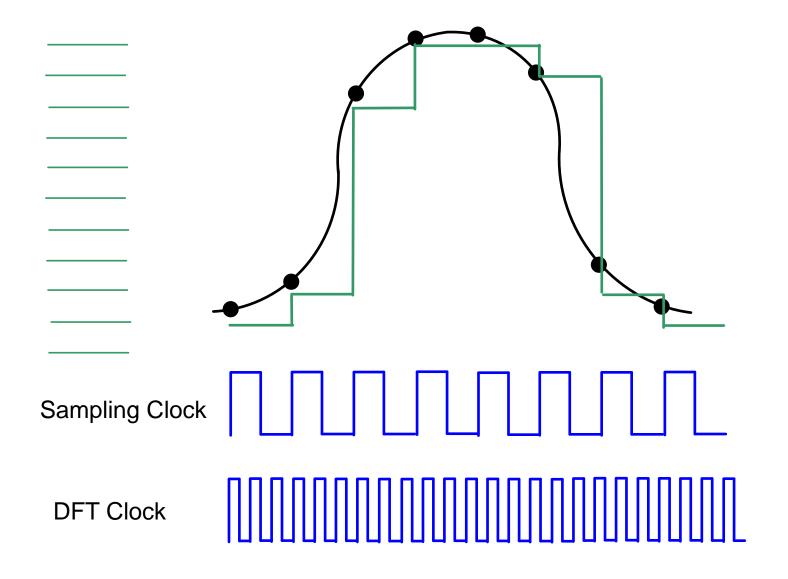


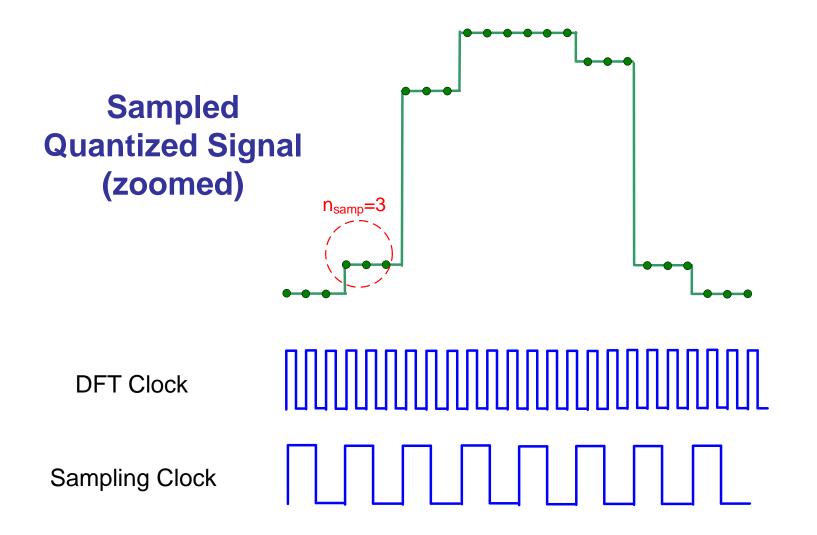
Quantized Sampled Input Signal (with zero-order sample and hold)



DFT Clock







Consider the following example

$$-f_{SIG}=50 Hz$$

$$-f_{CL}=500 \text{ Hz} \text{ (DAC clock)}$$

$$-f_{DFTCL}=71.24K$$
 Hz

(coherent sampling)

$$-n_{DFT}=15$$

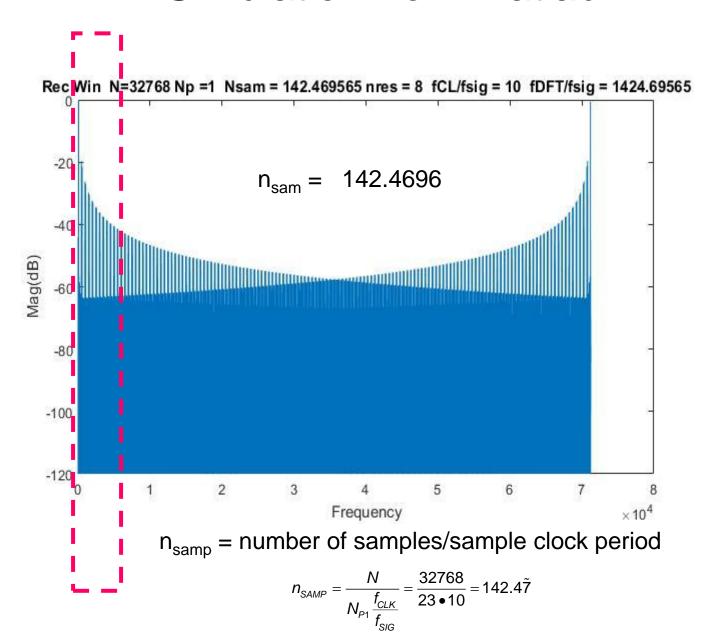
$$N = 2^{15} = 32,768$$

$$-N_{P1}=23$$
 (number of signal periods in DFT window)

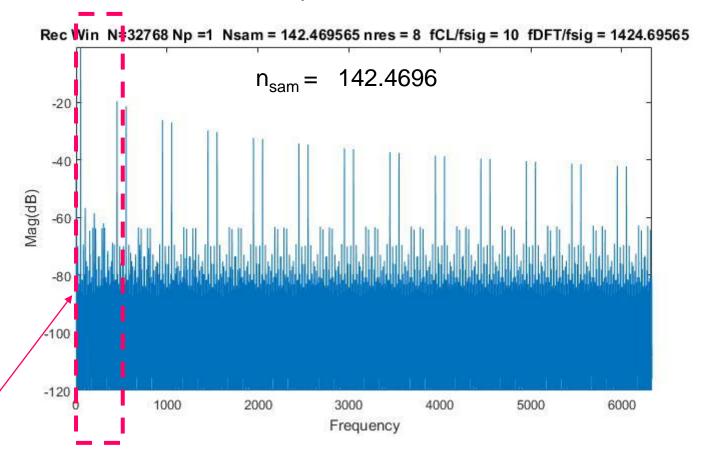
$$-N_P=1$$

$$- Xin(t) = .95sin(2\pi f_{SIG}t) (-.4455dB)$$

Matlab File: afft_Quantization_DAC_Jan2017.m

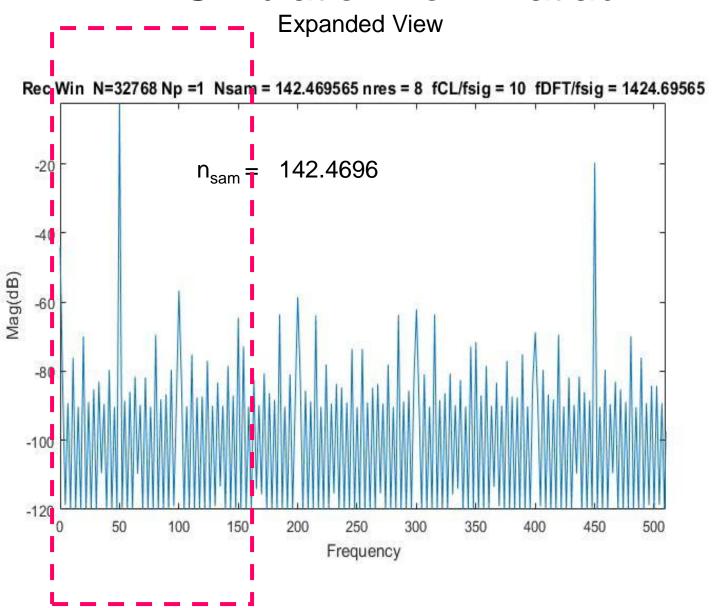


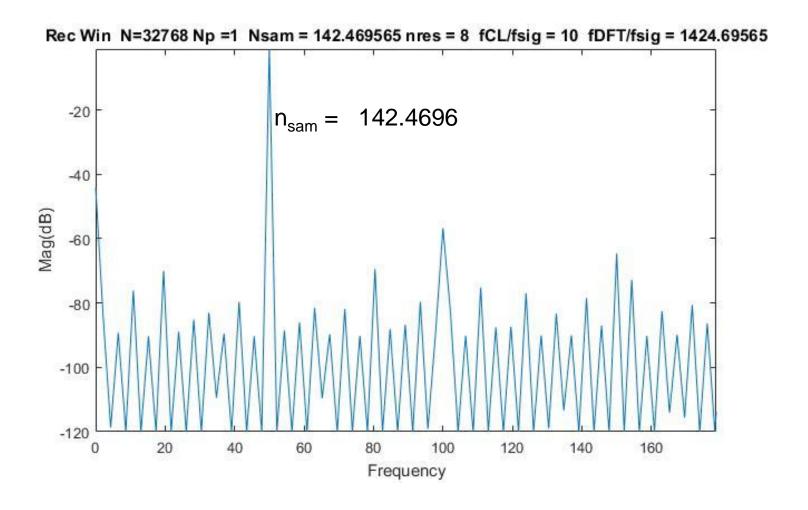
Expanded View



Width of this region is f_{CL}

Analogous to the overall DFT window when directly sampled but modestly asymmetric





DAC Comparisons with Quantization

Fundamental, second harmonic, and third harmonic

Columns 1 through 12	
-44.0825 -84.2069 -118.6751 -89.2265 -120.0000 -76.0893 -120.0000 -90.3321 -120.0000 -69.9163 -120.0000 -88.9097	
Columns 13 through 24	
-120.0000 -85.1896 -120.0000 -83.0183 -109.4722 -89.4980 -120.0000 -79.6110 -120.0000 -90.2992 -120.0000 -0.5960	
Columns 25 through 36	
-120.0000 -88.5446 -120.0000 -86.0169 -120.0000 -81.5409 -109.6386 -89.7275 -120.0000 -81.8340 -120.0000 -90.2331	
Columns 37 through 48	
-120.0000 -69.4356 -120.0000 -88.1400 -120.0000 -86.7214 -120.0000 -79.6273 -119.1428 -89.9175 -56.7024 -83.0511	
Columns 49 through 60	
-120.0000 -90.1331 -120.0000 -75.1821 -120.0000 -87.5706 -120.0000 -87.3205 -120.0000 -76.9769 -120.0000 -90.070	3
Columns 61 through 72	
-119.0588 -83.2950 -113.3964 -89.9982 -120.0000 -78.4288 -120.0000 -87.0328 -120.0000 -64.5409 -120.0000 -72.811	1
Columns 73 through 84	
-120.0000 -90.1876 -120.0000 -82.5616 -114.0867 -89.8269 -115.6476 -80.6553 -120.0000 -86.3818 -120.0000 -88.345	4
Columns 85 through 96	
-120.0000 -63.5207 -120.0000 -90.2704 -120.0000 -80.8524 -120.0000 -89.6174 -58.5435 -82.3253 -120.0000 -85.618	8

N	θ	Nsam	n	A ₁	A_2	A_3
32K	1	142.5	8	596	-56.7	-64.5
128K	1	569.9	8	596	-56.7	-64.45

(amplitude and time quantization)

Consider the following example

$$-f_{SIG}=50 Hz$$

$$-f_{CL}=500 \text{ Hz} \text{ (DAC clock)}$$

$$-f_{DFTCL}=71.24K$$
 Hz

(coherent sampling)

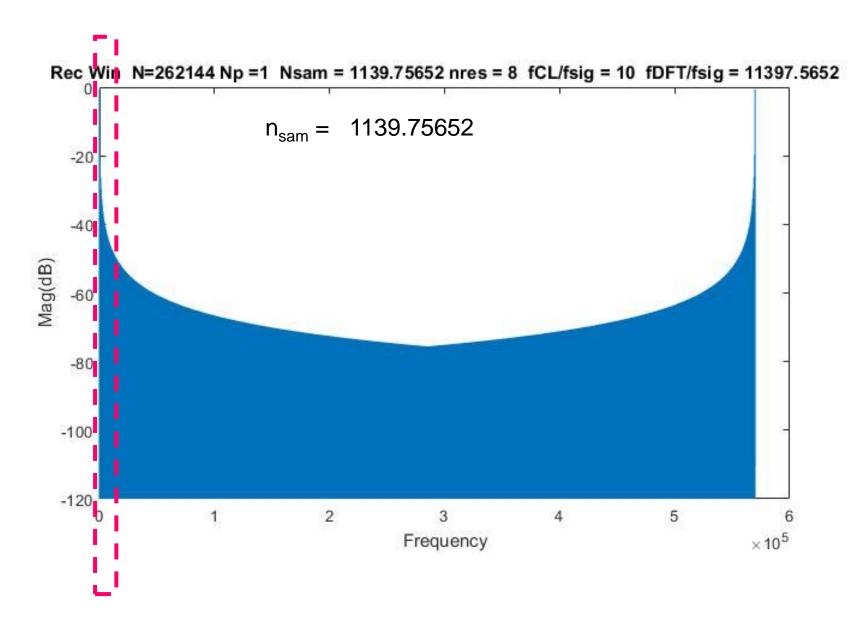
$$-n_{DFT} = 18$$

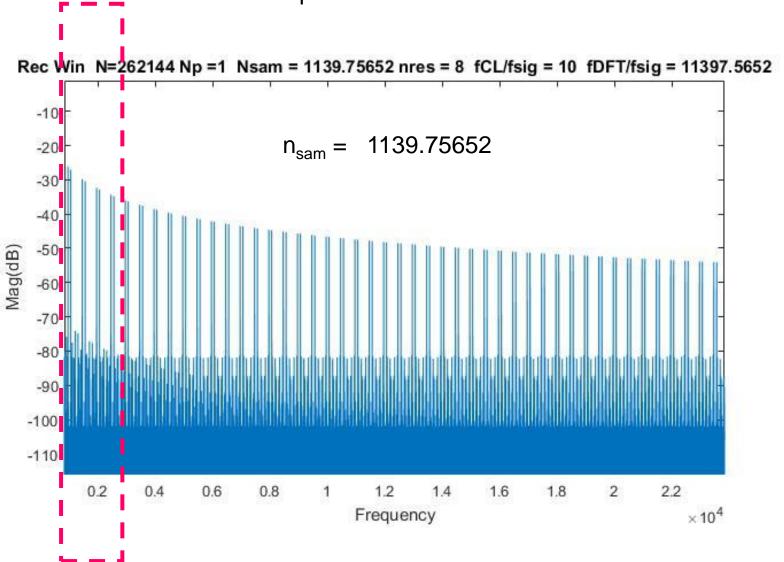
$$N = 2^{18} = 262,144$$

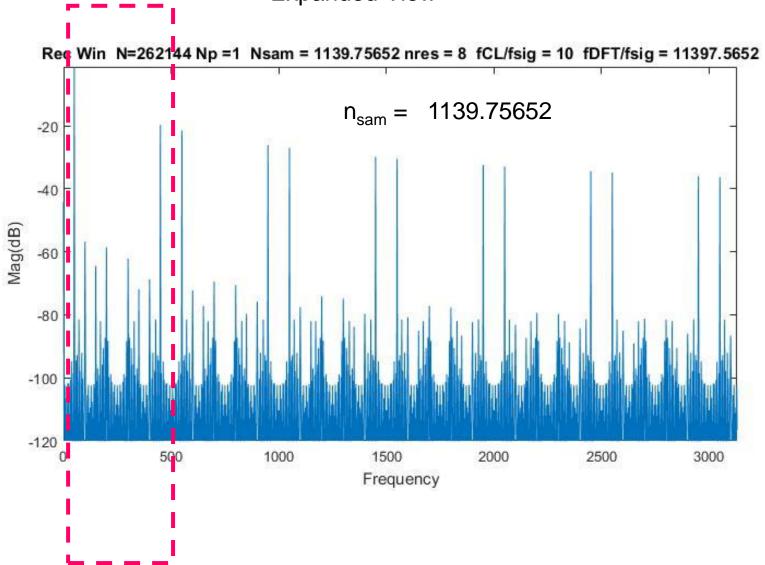
$$-N_{P1}=23$$
 (number of signal periods in DFT window)

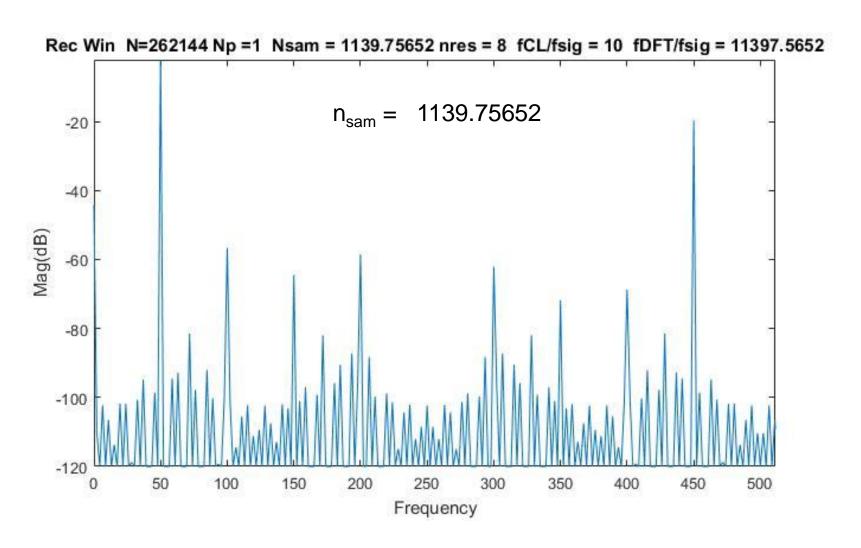
$$-N_P=1$$

$$- Xin(t) = .95sin(2\pi f_{SIG}t) (-.4455dB)$$









Consider the following example

$$-f_{SIG}=50 Hz$$

$$-f_{CL}$$
=500 Hz (DAC clock)

$$-f_{DFTCI} = 71.24K Hz$$
 (coherent sampling)

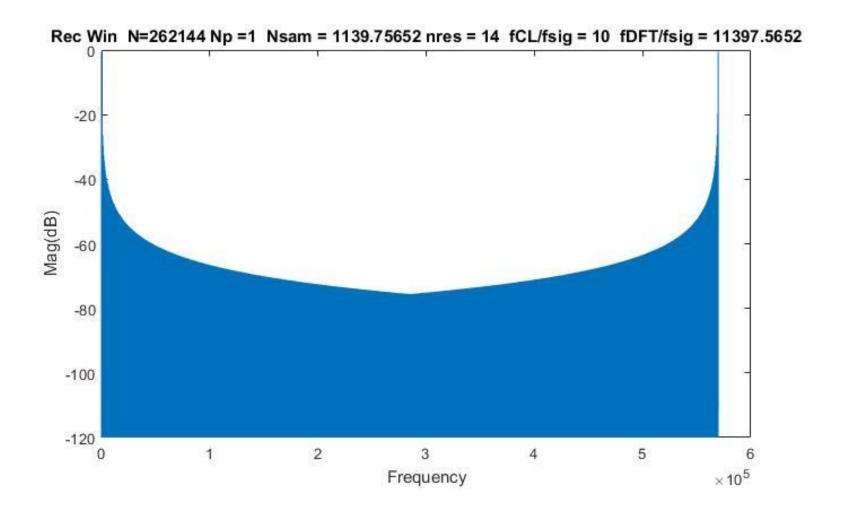
$$-n_{DFT}=18$$

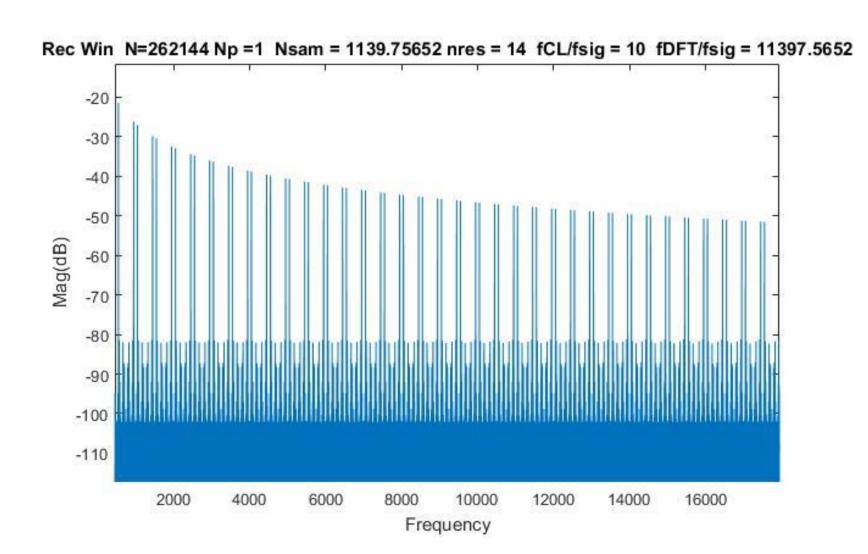
$$-N_{P1}=23$$
 (number of signal periods in DFT window)

$$-N_P=1$$

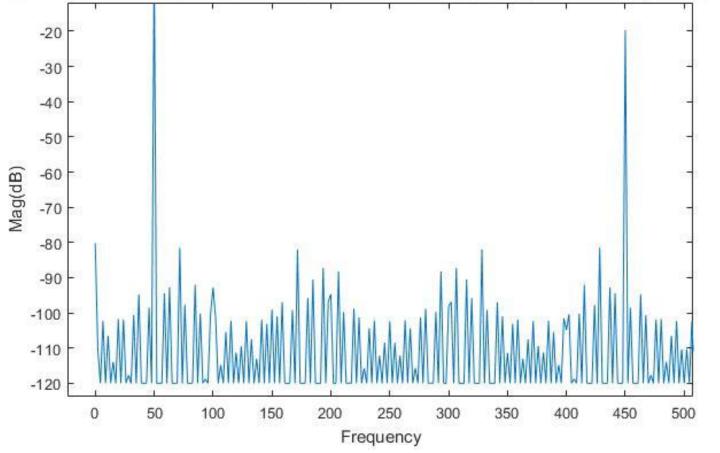
increased

$$- Xin(t) = .95sin(2\pi f_{SIG}t) (-.4455dB)$$





Rec Win N=262144 Np =1 Nsam = 1139.75652 nres = 14 fCL/fsig = 10 fDFT/fsig = 11397.5652



Consider the following example

$$-f_{SIG}=50 Hz$$

$$-f_{CL}=500 \text{ Hz} \text{ (DAC clock)}$$

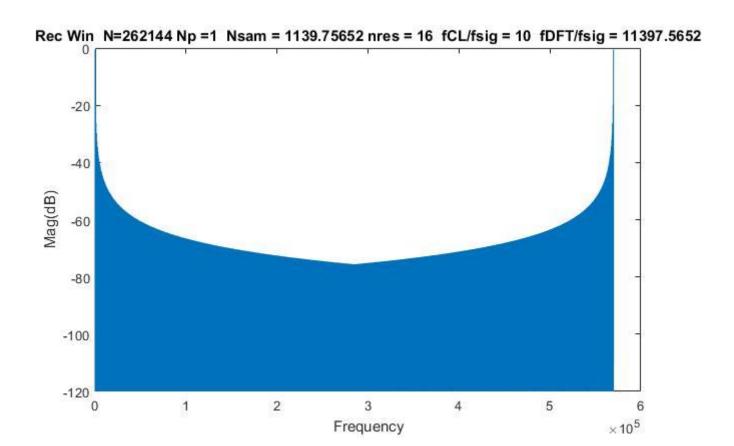
$$- n_{DFT} = 18$$

$$-N_{P1}=23$$
 (number of signal periods in DFT window)

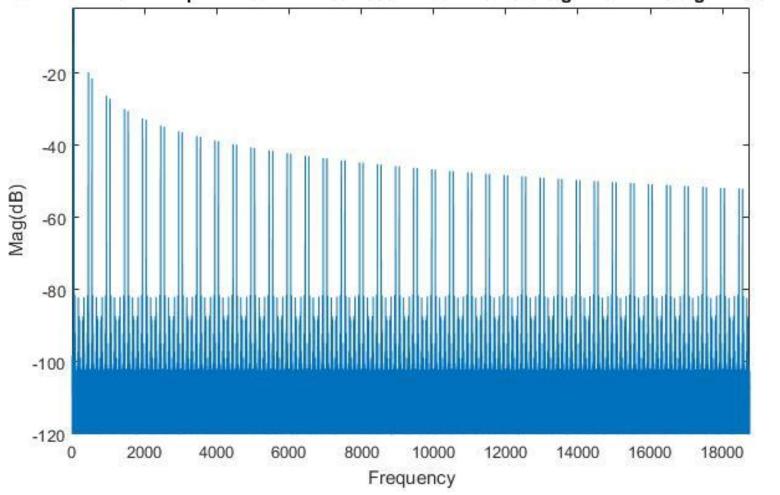
$$-N_{P}=1$$



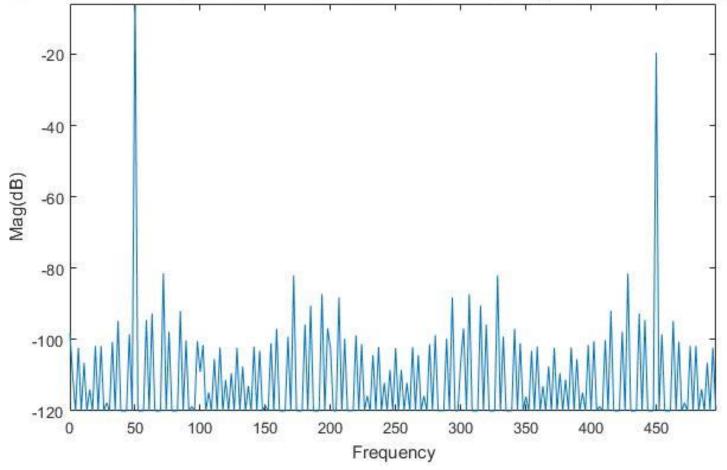
$$- Xin(t) = .95sin(2\pi f_{SIG}t) (-.4455dB)$$



Rec Win N=262144 Np =1 Nsam = 1139.75652 nres = 16 fCL/fsig = 10 fDFT/fsig = 11397.5652

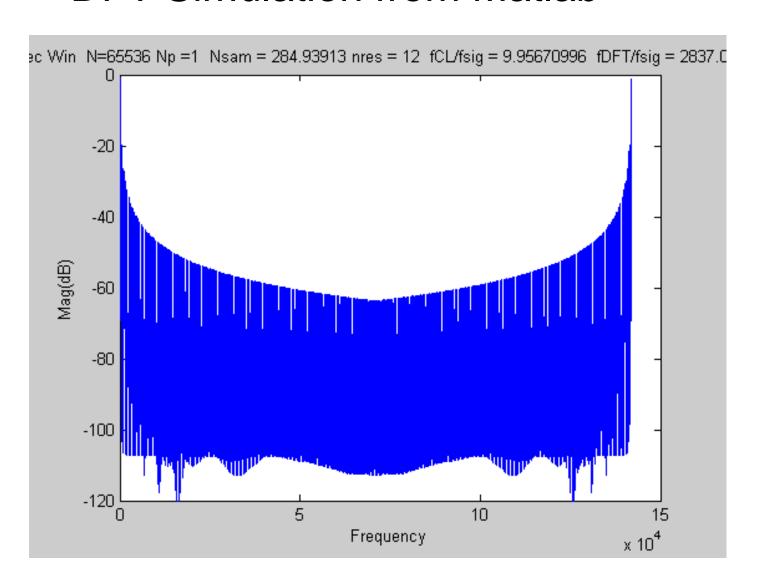


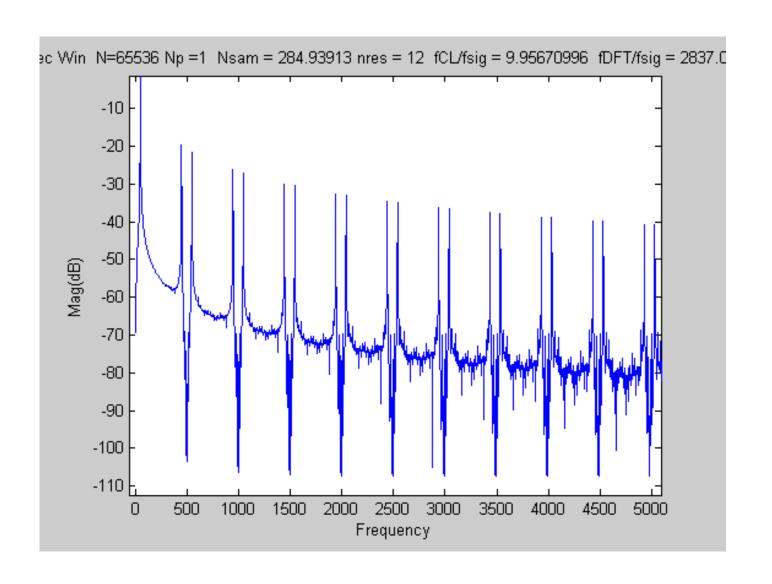
Rec Win N=262144 Np =1 Nsam = 1139.75652 nres = 16 fCL/fsig = 10 fDFT/fsig = 11397.5652

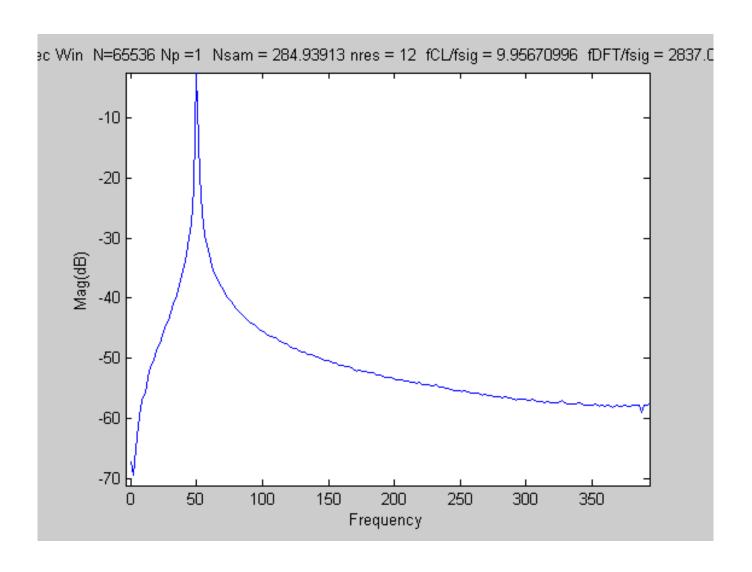


Consider the following example

- $-f_{SIG}=50 Hz$
- $-f_{CL}$ =497.8 Hz (DAC clock)
- $-f_{DFTCI} = 141.853K$ Hz (not coherent sampling)
- $-n_{DFT}=16$
- $-N_{P1}=23$ (number of signal periods in DFT window)
- $-N_P=1$
- n_{res}=16bits
- $-Xin(t) = .95sin(2\pi f_{SIG}t)$ (-.4455dB)







Summary of time and amplitude quantization assessment

Time and amplitude quantization do not introduce <u>harmonic</u> distortion

Time and amplitude quantization do increase the noise floor



Stay Safe and Stay Healthy!

End of Lecture 7